

LM4985

*LM4985 Stereo 135mW Low Noise Headphone Amplifier with Selectable
Capacitively Coupled or Output Capacitor-less (OCL) Output and Digitally
Controlled (I2C) Volume Control*



Literature Number: SNAS346B

LM4985 Boomer® Audio Power Amplifier Series

Stereo 135mW Low Noise Headphone Amplifier with Selectable Capacitively Coupled or Output Capacitor-less (OCL) Output and Digitally Controlled (I²C) Volume Control

General Description

The LM4985 is a stereo audio power amplifier with internal digitally controlled volume control. This amplifier is capable of delivering 68mW_{RMS} per channel into a 16Ω load or 38mW_{RMS} per channel into a 32Ω load at 1% THD when powered by a 3.6V power supply and operating in the OCL mode.

Boomer audio power amplifiers were designed specifically to provide high quality output power with a minimal amount of external components. To that end, the LM4985 features two functions that optimize system cost and minimize PCB area: an integrated, digitally controlled (I²C bus) volume control and an operational mode that eliminates output signal coupling capacitors (OCL mode). Since the LM4985 does not require bootstrap capacitors, snubber networks, or output coupling capacitors, it is optimally suited for low-power, battery powered portable systems. For added design flexibility, the LM4985 can also be configured for single-ended capacitively coupled outputs.

The LM4985 features a current shutdown mode for micropower dissipation and thermal shutdown protection.

Key Specifications (V_{DD} = 3.6V)

■ PSRR: 217Hz and 1kHz	
Output Capacitor-less (OCL)	
f _{RIPPLE} = 217Hz	77dB (typ)
f _{RIPPLE} = 1kHz	76dB (typ)
Capacitor Coupled (C-CUPL)	
f _{RIPPLE} = 217Hz	63dB (typ)
f _{RIPPLE} = 1kHz	62dB (typ)
■ Output Power per channel (f _{IN} = 1kHz, THD+N = 1%), R _L = 16Ω, OCL	
V _{DD} = 2.5V	31mW (typ)
V _{DD} = 3.6V	68mW (typ)
V _{DD} = 5.0V	135mW (typ)
■ THD+N (f = 1kHz)	
R _{LOAD} = 16Ω, OCL, P _{OUT} = 60mW	0.60
R _{LOAD} = 32Ω, OCL, P _{OUT} = 33mW	0.031
■ Shutdown Current	0.1μA (typ)

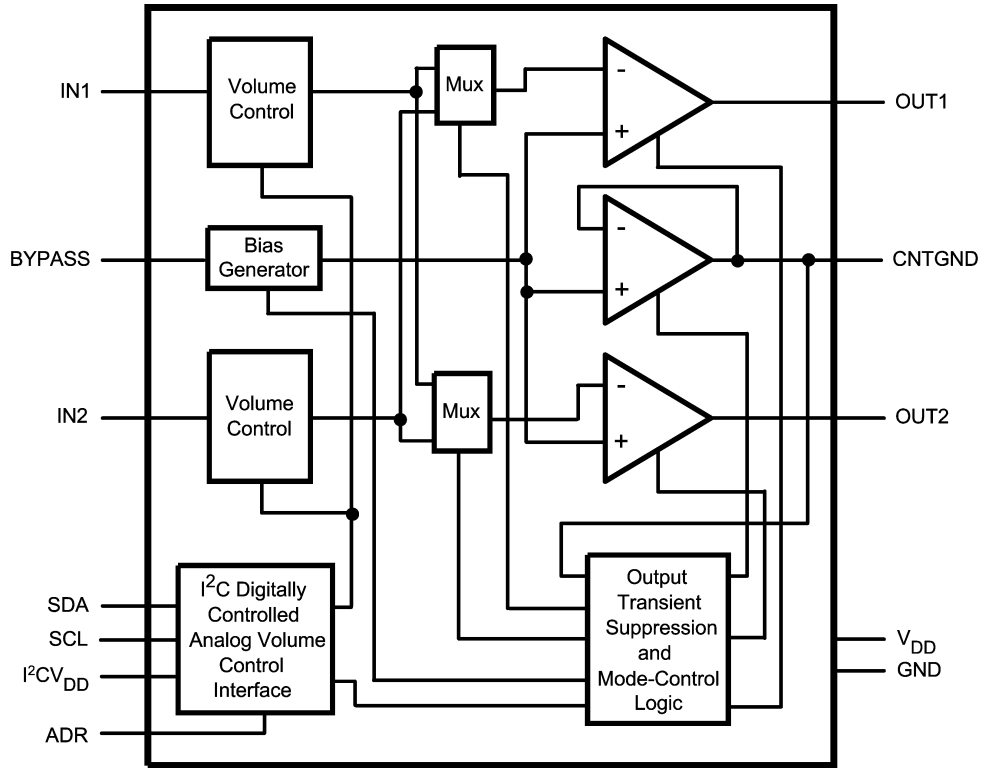
Features

- OCL or capacitively coupled outputs (patent pending)
- I²C Digitally Controlled Volume Control
- Available in space-saving 0.4mm lead-pitch micro SMD package
- Volume control range: -76dB to +18dB
- Ultra low current shutdown mode
- 2.3V - 5.5V operation
- Ultra low noise

Applications

- Mobile Phones
- PDAs
- Portable electronics devices
- MP3 Players

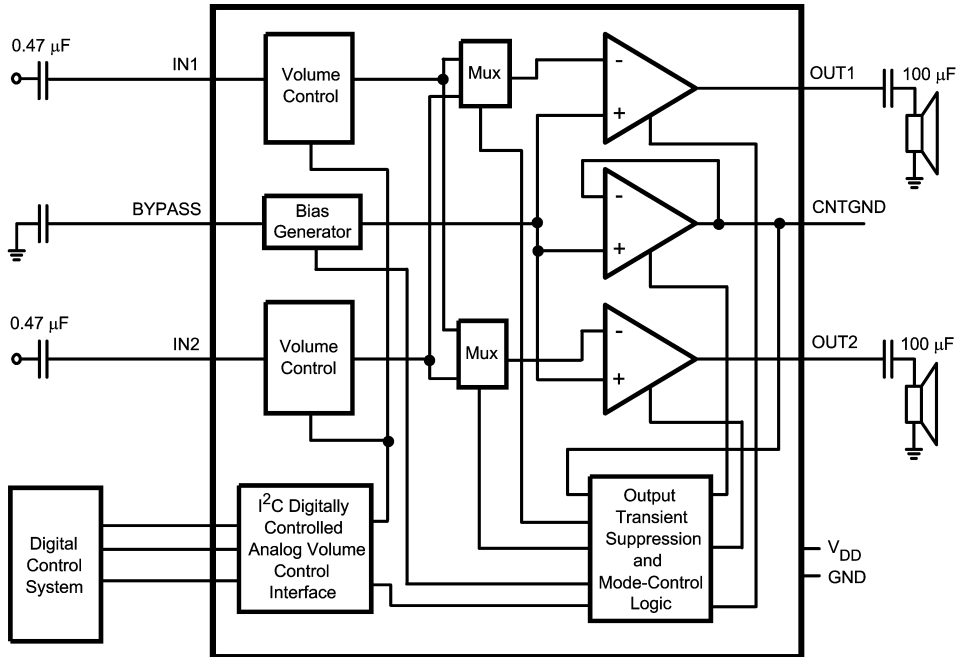
Block Diagram



201697E9

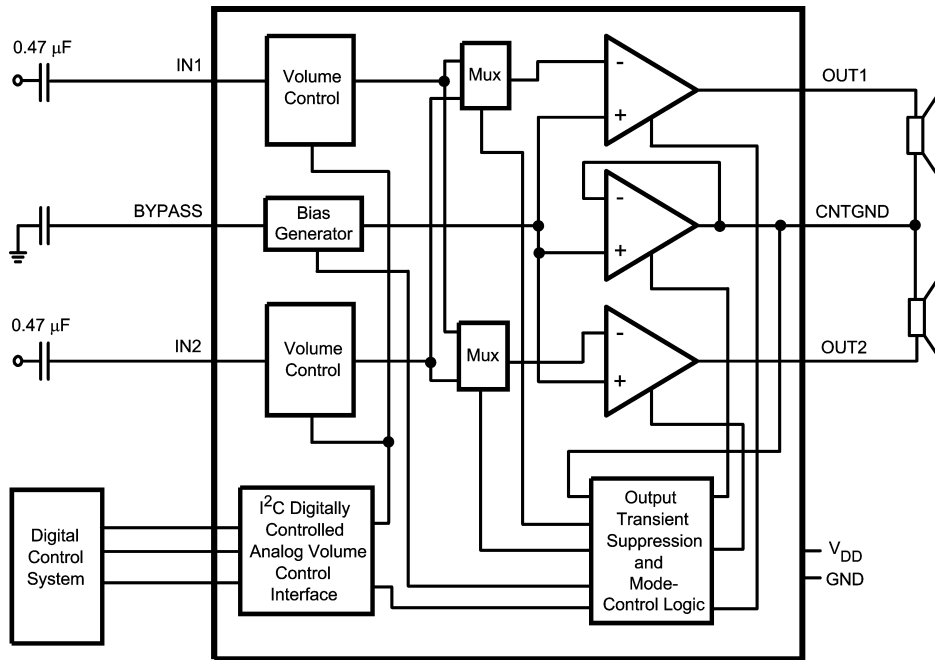
FIGURE 1. Block Diagram

Typical Application



201697F0

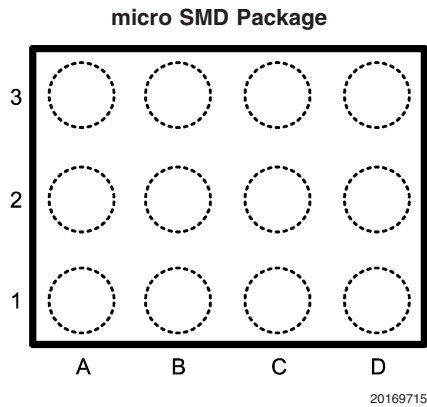
FIGURE 2. Typical Capacitively Coupled Output Configuration Circuit



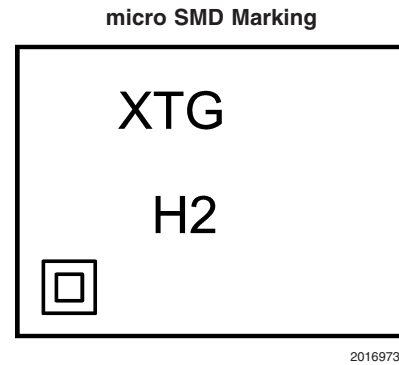
201697F1

FIGURE 3. Typical OCL Output Configuration Circuit

Connection Diagrams



Top View
 Order Number LM4985TM
 See NS Package Number TMD12AAA



Top View
 X – Date Code
 T – Die Traceability
 G – Boomer Family
 H2 – LM4985TM

Pin Reference, Name, and Function

Reference	Name	Function
A1	ADR	I ² C serial interface address input.
A2	IN2	Analog signal input two.
A3	OUT2	Power amplifier two output.
B1	SDA	I ² C serial interface data input.
B2	BYPASS	The internal $V_{DD}/2$ ac bypass node.
B3	CNTGND	In OCL mode, this is the ac ground return. It is biased to $V_{DD}/2$. Leave unconnected for C-CUPL mode.
C1	SCL	I ² C serial interface clock input.
C2	GND	The LM4985's power supply ground input.
C3	V_{DD}	The LM4985's power supply voltage input.
D1	I ² CV _{DD}	I ² C serial interface power supply input. Can be connected to the same supply that is connected to the V_{DD} pin.
D2	IN1	Analog signal input one.
D3	OUT1	Power amplifier one output.

Absolute Maximum Ratings (Notes 1, 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage (V_{DD} , I^2CV_{DD})	6.0V
Storage Temperature	-65°C to +150°C
Input Voltage (IN1, IN2, OUT1, OUT2, BYPASS, CNTGND, GND pins relative to the V_{DD} pin)	-0.3V to $V_{DD} + 0.3V$
Input Voltage (ADR, SDA, SCL pins, relative to the I^2CV_{DD} pin)	-0.3V to $I^2CV_{DD} + 0.3V$
Power Dissipation (Note 3)	Internally Limited
ESD Susceptibility (Note 4)	2000V

ESD Susceptibility (Note 5)	200V
Junction Temperature	150°C
Thermal Resistance	
θ_{JA}	109°C/W

Operating Ratings

Temperature Range	$T_{MIN} \leq T_A \leq T_{MAX}$	-40°C $\leq T_A \leq$ 85°C
Supply Voltage	V_{DD}	$2.3V \leq V_{CC} \leq 5.5V$
	I^2CV_{DD}	$1.7V \leq I^2CV_{DD} \leq 5.5V$

Electrical Characteristics $V_{DD} = 5V$ (Notes 1, 2)

The following specifications apply for $R_L = 16\Omega$, $f = 1kHz$, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^\circ C$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)
			Typ (Note 6)	Limit (Notes 7, 8)	
I_{DD}	Quiescent Power Supply Current	$V_{IN} = 0V$, $I_{OUT} = 0A$	2		mA (max)
		Single-Channel no load OCL	1.5		
		Single-Channel no load C-CUPL	3	4.9	
		Dual-Channel no load OCL	2.3	3.8	
I_{SD}	Shutdown Current	$V_{SHUTDOWN} = GND$	0.1	1.0	μA (max)
V_{SDIH}	Logic Voltage Input High			3.5	V (min)
V_{SDIL}	Logic Voltage Input Low			1.5	V (max)
P_O	Output Power	THD $\leq 1\%$; $f_{IN} = 1kHz$			mW (min)
		$R_{LOAD} = 16\Omega$ OCL	135	115	
		$R_{LOAD} = 16\Omega$ C-CUPL	135		
		$R_{LOAD} = 32\Omega$ OCL	79	70	
THD+N	Total Harmonic Distortion + Noise	$R_{LOAD} = 16\Omega$ OCL, $P_O = 100mW$	0.08		%
		$R_{LOAD} = 16\Omega$ C-CUPL, $P_O = 100mW$	0.02		
		$R_{LOAD} = 32\Omega$ OCL, $P_O = 60mW$	0.04		
		$R_{LOAD} = 32\Omega$ C-CUPL, $P_O = 70mW$	0.01		
V_{ON}	Output Noise Voltage	$V_{IN} = AC$ GND, $A_V = 0dB$, A-weighted	15		μV
PSRR	Power Supply Rejection Ratio	$V_{RIPPLE} = 200mV_{p-p}$ (Note 9)			dB (min)
		$f_{IN} = 217Hz$ sinewave			
		OCL	77	57	
		C-CUPL	65		
Xtalk	Channel-to-channel Crosstalk	$f_{IN} = 1kHz$ sinewave			dB
		OCL	77	60	
		C-CUPL	65		
		$P_{out} = 40mW$. OCL			
Xtalk	Channel-to-channel Crosstalk	$R_{LOAD} = 16\Omega$	51		dB
		$R_{LOAD} = 32\Omega$	56		
		$P_{out} = 50mW$. C-CUPL			dB
		$R_{LOAD} = 16\Omega$	58		
	$R_{LOAD} = 32\Omega$	68			

Electrical Characteristics $V_{DD} = 5V$ (Notes 1, 2) (Continued)

The following specifications apply for $R_L = 16\Omega$, $f = 1\text{kHz}$, and $C_B = 4.7\mu\text{F}$ unless otherwise specified. Limits apply to $T_A = 25^\circ\text{C}$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)
			Typ (Note 6)	Limit (Notes 7, 8)	
T_{WU}	Wake Up Time form Shutdown	$C_{BYPASS} = 4.7\mu\text{F}$ (Note 11)			msec
		WT1 = 0, WT0 = 0 OCL C-CUPL	75 285		
		WT1 = 0, WT0 = 1 OCL C-CUPL	110 530		
		WT1 = 1, WT0 = 0 OCL C-CUPL	180 1030		
		WT1 = 1, WT0 = 1 OCL C-CUPL	320 2050		
R_{IN}	Input Resistance	Stereo mode	20		k Ω
		Mono mode	10		
A_{VMIN}	Minimum Gain	Code = 00000	-76		dB (min)
A_{VMAX}	Maximum Gain	Code = 11111	18		dB (min)
ΔA_V	Gain Accuracy per Step	$18\text{dB} \geq A_V \geq -44\text{dB}$ $-44\text{dB} \geq A_V > -76\text{dB}$	± 0.5 ± 1.0		dB
V_{OS}	Output Offset Voltage	OCL $R_{LOAD} = 32\Omega$ $V_{IN} = \text{AC GND}$	2.0	20	mV (max)

Electrical Characteristics $V_{DD} = 3.6V$ (Notes 1, 2)

The following specifications apply for $R_L = 16\Omega$, $f = 1\text{kHz}$, and $C_B = 4.7\mu\text{F}$ unless otherwise specified. Limits apply to $T_A = 25^\circ\text{C}$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)
			Typ (Note 6)	Limit (Notes 7, 8)	
I_{DD}	Quiescent Power Supply Current	$V_{IN} = 0V$, $I_{OUT} = 0A$ Single-Channel no load OCL	1.8	3.1	mA (max)
		Single-Channel no load C-CUPL	1.0		
		Dual-Channel no load OCL	2.1	4	
		Dual-Channel no load C-CUPL	2.3	3	
I_{SD}	Shutdown Current	$V_{SHUTDOWN} = \text{GND}$	0.1	1.0	μA (max)
V_{SDIH}	Logic Voltage Input High			2.52	V (min)
V_{SDIL}	Logic Voltage Input Low			1.08	V (max)
P_O	Output Power	THD+N < 1%, $f_{IN} = 1\text{kHz}$ $R_{LOAD} = 16\Omega$ OCL	68	60	mW (min)
		$R_{LOAD} = 16\Omega$ C-CUPL	70		
		$R_{LOAD} = 32\Omega$ OCL	38	34	
		$R_{LOAD} = 32\Omega$ C-CUPL	41		

Electrical Characteristics $V_{DD} = 3.6V$ (Notes 1, 2) (Continued)

The following specifications apply for $R_L = 16\Omega$, $f = 1kHz$, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^\circ C$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)
			Typ (Note 6)	Limit (Notes 7, 8)	
THD+N	Total Harmonic Distortion + Noise	$R_{LOAD} = 16\Omega$ OCL, $P_O = 60mW$	0.06		%
		$R_{LOAD} = 16\Omega$ C-CUPL, $P_O = 60mW$	0.03		
		$R_{LOAD} = 32\Omega$ OCL, $P_O = 33mW$	0.03		
		$R_{LOAD} = 32\Omega$ C-CUPL, $P_O = 38mW$	0.03		
V_{ON}	Output Noise Voltage	$V_{IN} = AC$ GND, $A_V = 0dB$, A-weighted	15		μV
PSRR	Power Supply Rejection Ratio	$V_{RIPPLE} = 200mVp-p$ (Note 9)			dB (min)
		$f_{IN} = 217Hz$ sinewave			
		OCL	77	55	
		C-CUPL	63		
Xtalk	Channel-to-Channel Crosstalk	$P_{out} = 40mW$. OCL			dB
		$R_{LOAD} = 16\Omega$	51		
		$R_{LOAD} = 32\Omega$	56		
		$P_{out} = 50mW$. C-CUPL			dB
$R_{LOAD} = 16\Omega$	58				
$R_{LOAD} = 32\Omega$	69				
T_{WU}	Wake Up Time from Shutdown	$C_{BYPASS} = 4.7\mu F$ (Note 11)			msec
		WT1 = 0, WT0 = 0			
		OCL	66	93	
		C-CUPL	222		
		WT1 = 0, WT0 = 1			
		OCL	92		
		C-CUPL	405		
		WT1 = 1, WT0 = 0			
OCL	143				
C-CUPL	774				
R_{IN}	Input Resistance	Stereo mode	20		$k\Omega$
		Mono mode	10		
A_{VMIN}	Minimum Gain	Code = 00000	-76	-72	dB (max)
A_{VMAX}	Maximum Gain	Code = 11111	18	17	dB (min)
ΔA_V	Gain Accuracy per Step	$18dB \geq A_V \geq -44dB$ $-44dB \geq A_V > -76dB$	± 0.5 ± 1.0	± 1.0 ± 2.0	dB
V_{OS}	Output Offset Voltage	OCL $R_{LOAD} = 32\Omega$ $V_{IN} = AC$ GND	2.0	20	mV (max)

Electrical Characteristics $V_{DD} = 2.5V$ (Notes 1, 2)

The following specifications apply for $R_L = 16\Omega$, $f = 1kHz$, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^\circ C$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)
			Typ (Note 6)	Limit (Notes 7, 8)	
I_{DD}	Quiescent Power Supply Current	$V_{IN} = 0V, I_{OUT} = 0A$ Single-Channel no load OCL	1.6		mA
		Single-Channel no load C-CUPL	1		
		Dual-Channel no load OCL	2.1		
		Dual-Channel no load C-CUPL	1.6		
I_{SD}	Shutdown Current	$V_{SHUTDOWN} = GND$	0.1		μA
V_{SDIH}	Logic Voltage Input High			1.75	V (min)
V_{SDIL}	Logic Voltage Input Low			0.75	V (max)
P_O	Output Power	THD+N < 1%, $f_{IN} = 1kHz$ $R_{LOAD} = 16\Omega$ OCL	31		mW
		$R_{LOAD} = 16\Omega$ C-CUPL	33		
		$R_{LOAD} = 32\Omega$ OCL	19		
		$R_{LOAD} = 32\Omega$ C-CUPL	19		
THD+N	Total Harmonic Distortion + Noise	$R_{LOAD} = 16\Omega$ OCL, $P_O = 26mW$	0.07		%
		$R_{LOAD} = 16\Omega$ C-CUPL, $P_O = 20mW$	0.05		
		$R_{LOAD} = 32\Omega$ OCL, $P_O = 16mW$	0.06		
		$R_{LOAD} = 32\Omega$ C-CUPL, $P_O = 15mW$	0.04		
V_{ON}	Output Noise Voltage	$V_{IN} = AC$ GND, $A_V = 0dB$, A-weighted	10		μV
PSRR	Power Supply Rejection Ratio	$V_{RIPPLE} = 200mVp-p$ (Note 9) $f_{IN} = 217Hz$ sinewave OCL	75		dB
		C-CUPL	59		
		$f_{IN} = 1kHz$ sinewave OCL	75		
		C-CUPL	59		
Xtalk	Channel-to-Channel Crosstalk	$P_{out} = 20mW$, OCL $R_{LOAD} = 16\Omega$	50		dB
		$R_{LOAD} = 32\Omega$	55		
		$P_{out} = 20mW$, C-CUPL $R_{LOAD} = 16\Omega$	58		dB
		$R_{LOAD} = 32\Omega$	67		
T_{WU}	Wake Up Time from Shutdown	$C_{BYPASS} = 4.7\mu F$ (Note 11) WT1 = 0, WT0 = 0 OCL	66		msec
		C-CUPL	214		
		WT1 = 0, WT0 = 1 OCL	92		
		C-CUPL	544		
		WT1 = 1, WT0 = 0 OCL	145		
		C-CUPL	1053		
		WT1 = 1, WT0 = 1 OCL	250		
		C-CUPL	2098		
R_{IN}	Input Resistance	Stereo mode	20		$k\Omega$
		Mono mode	10		
A_{VMIN}	Minimum Gain	Code = 00000	-76		dB

Electrical Characteristics $V_{DD} = 2.5V$ (Notes 1, 2) (Continued)

The following specifications apply for $R_L = 16\Omega$, $f = 1kHz$, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^\circ C$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)
			Typ (Note 6)	Limit (Notes 7, 8)	
A_{VMAX}	Maximum Gain	Code = 11111	18		dB
ΔA_V	Gain Accuracy per Step	$18dB \geq A_V \geq -44dB$ $-44dB \geq A_V > -76dB$	± 0.5 ± 1.0		dB
V_{OS}	Output Offset Voltage	OCL $R_{LOAD} = 32\Omega$ $V_{IN} = AC\ GND$	2.0		mV

Note 1: All voltages are measured with respect to the GND pin unless otherwise specified.

Note 2: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not guarantee specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which guarantee specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not guaranteed for parameters where no limit is given, however, the typical value is a good indication of device performance.

Note 3: The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature, T_A . The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$ or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4985, see power derating currents for more information.

Note 4: Human Body Model: 100pF discharged through a 1.5k Ω resistor.

Note 5: Machine Model: 200pF $\leq C_{mm} \leq 220pF$ discharged through all pins.

Note 6: Typical values are measured at 25°C and represent the parametric norm.

Note 7: Limits are guaranteed to National's AOQL (Average Outgoing Quality Level).

Note 8: Datasheet min/max specification limits are guaranteed by design, test, or statistical analysis.

Note 9: 10 Ω terminated input.

Note 10: The LDA10A package has its exposed-DAP soldered to an exposed 1.2in² area of 1oz. Printed circuit board copper.

Note 11: The wake-up time (T_{WU}) is calculated using the following formula; $T_{WU} = [C_{BYPASS} (VDD) / 2 (I_{BYPASS})] + 40ms$.

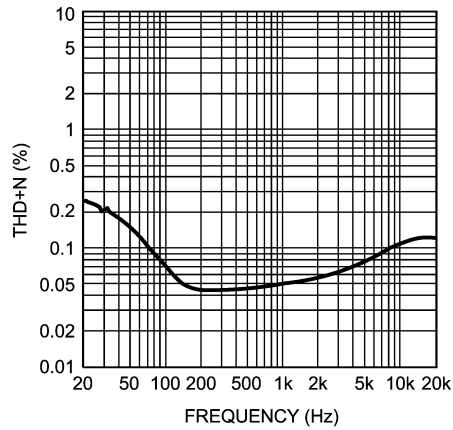
External Components Description (Figure 2)

Components		Functional Description
1.	C_i	Input coupling capacitor which blocks the DC voltage at the amplifier's input terminals. Also creates a high-pass filter with R_i at $f_c = 1/(2\pi R_i C_i)$. Refer to the section Proper Selection of External Components , for an explanation of how to determine the value of C_i .
2.	C_S	Supply bypass capacitor which provides power supply filtering. Refer to the Power Supply Bypassing section for information concerning proper placement and selection of the supply bypass capacitor.
3.	C_B	Bypass pin capacitor which provides half-supply filtering. Refer to the section, Proper Selection of Proper Components , for information concerning proper placement and selection of C_B .
6.	C_o	Output coupling capacitor which blocks the DC voltage at the amplifier's output. Forms a high pass filter with R_L at $f_o = 1/(2\pi R_L C_o)$.

Typical Performance Characteristics

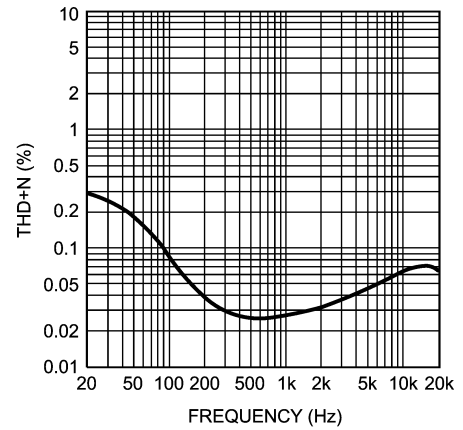
$T_A = 25^\circ\text{C}$, $A_V = 0\text{dB}$, $f_{IN} = 1\text{kHz}$ unless otherwise stated.

THD+N vs Frequency
 $V_{DD} = 2.5\text{V}$, $R_L = 16\Omega$
 $P_{OUT} = 20\text{mW}$, C-CUPL



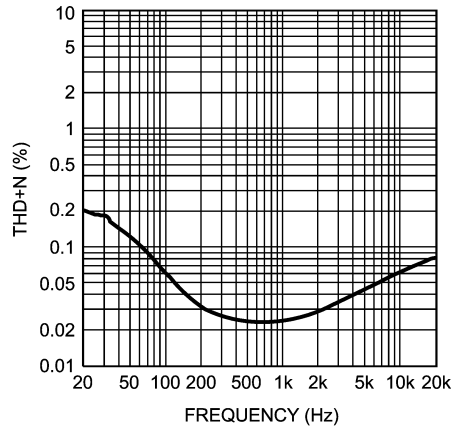
201697D1

THD+N vs Frequency
 $V_{DD} = 3.6\text{V}$, $R_L = 16\Omega$
 $P_{OUT} = 50\text{mW}$, C-CUPL



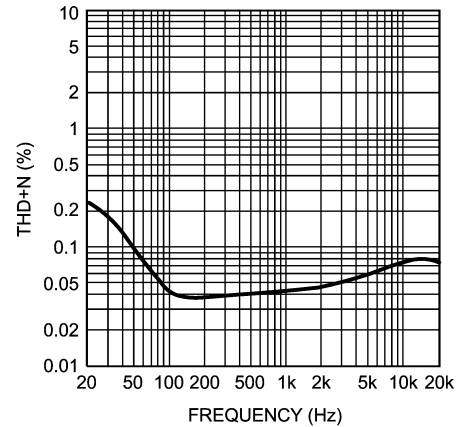
201697D2

THD+N vs Frequency
 $V_{DD} = 5\text{V}$, $R_L = 16\Omega$
 $P_{OUT} = 50\text{mW}$, C-CUPL



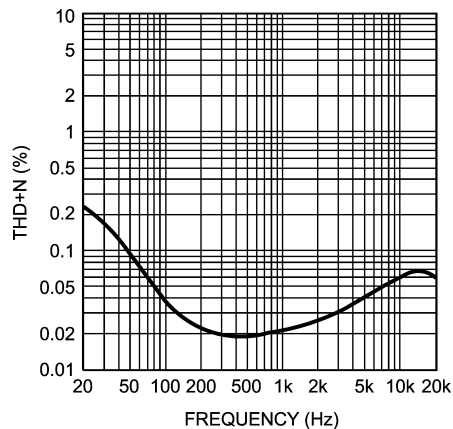
201697D3

THD+N vs Frequency
 $V_{DD} = 2.5\text{V}$, $R_L = 32\Omega$
 $P_{OUT} = 15\text{mW}$, C-CUPL



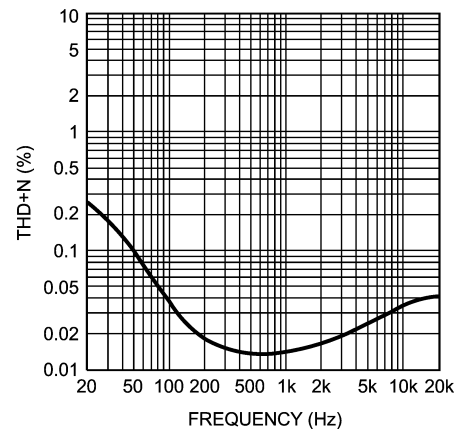
201697D4

THD+N vs Frequency
 $V_{DD} = 3.6\text{V}$, $R_L = 32\Omega$
 $P_{OUT} = 35\text{mW}$, C-CUPL



201697D5

THD+N vs Frequency
 $V_{DD} = 5.0\text{V}$, $R_L = 32\Omega$
 $P_{OUT} = 60\text{mW}$, C-CUPL

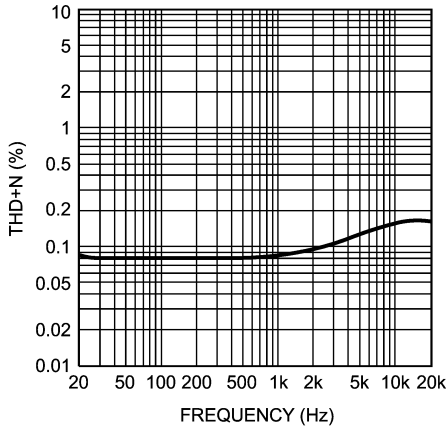


201697D6

Typical Performance Characteristics (Continued)

THD+N vs Frequency

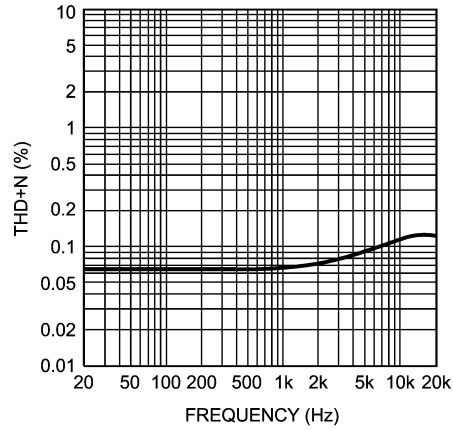
$V_{DD} = 2.5V$, $R_L = 16\Omega$
 $P_{OUT} = 20mW$, OCL



20169764

THD+N vs Frequency

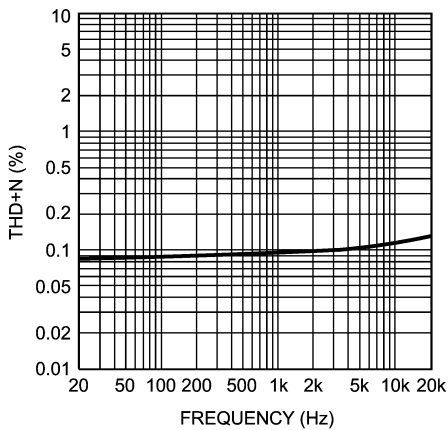
$V_{DD} = 3.6V$, $R_L = 16\Omega$
 $P_{OUT} = 50mW$, OCL



20169765

THD+N vs Frequency

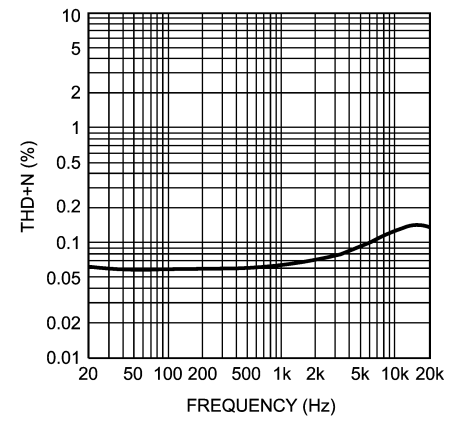
$V_{DD} = 5.0V$, $R_L = 16\Omega$
 $P_{OUT} = 50mW$, OCL



20169766

THD+N vs Frequency

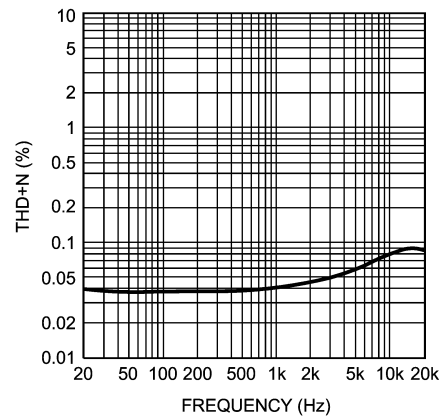
$V_{DD} = 2.5V$, $R_L = 32\Omega$
 $P_{OUT} = 15mW$, OCL



20169767

THD+N vs Frequency

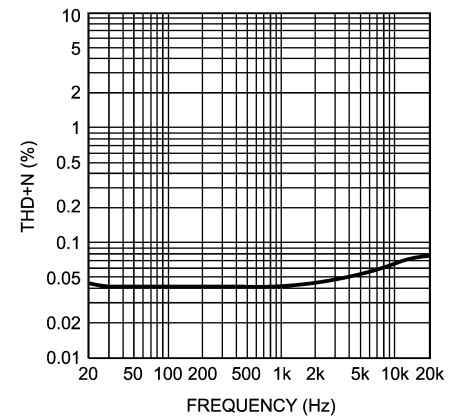
$V_{DD} = 3.6V$, $R_L = 32\Omega$
 $P_{OUT} = 35mW$, OCL



20169768

THD+N vs Frequency

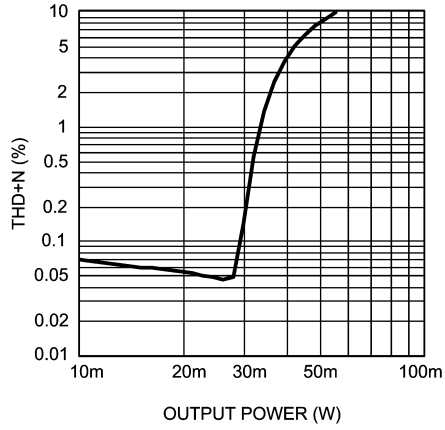
$V_{DD} = 5.0V$, $R_L = 32\Omega$
 $P_{OUT} = 60mW$, OCL



20169769

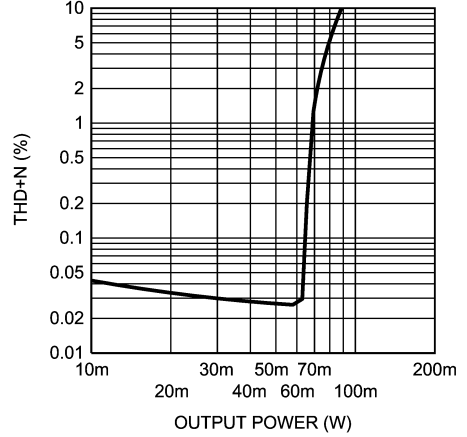
Typical Performance Characteristics (Continued)

THD+N vs Output Power
 $V_{DD} = 2.5V, R_L = 16\Omega$
 C-CUPL



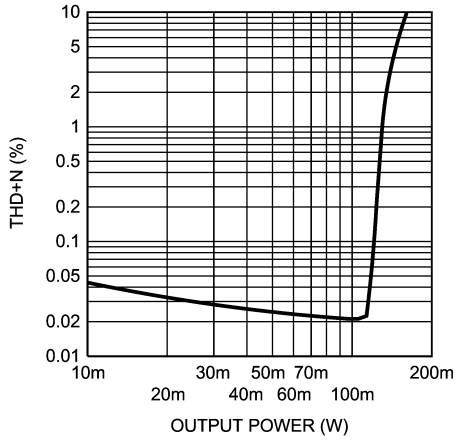
201697H3

THD+N vs Output Power
 $V_{DD} = 3.6V, R_L = 16\Omega$
 C-CUPL



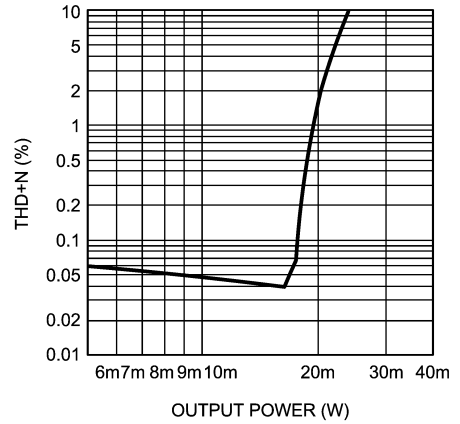
201697C6

THD+N vs Output Power
 $V_{DD} = 5.0V, R_L = 16\Omega$
 C-CUPL



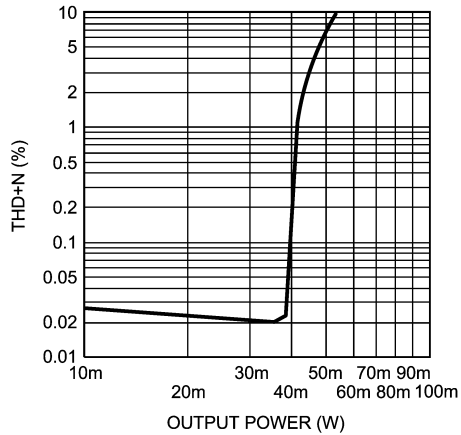
201697C7

THD+N vs Output Power
 $V_{DD} = 2.5V, R_L = 32\Omega$
 C-CUPL



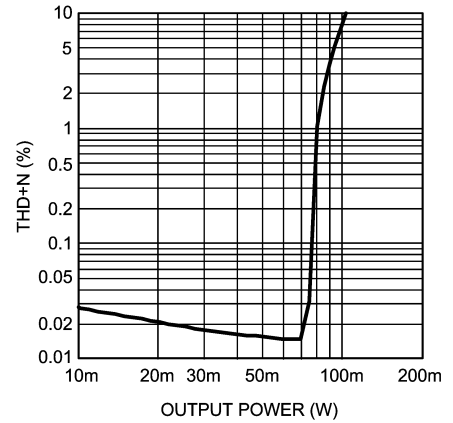
201697F2

THD+N vs Output Power
 $V_{DD} = 3.6V, R_L = 32\Omega$
 C-CUPL



201697F3

THD+N vs Output Power
 $V_{DD} = 5.0V, R_L = 32\Omega$
 C-CUPL

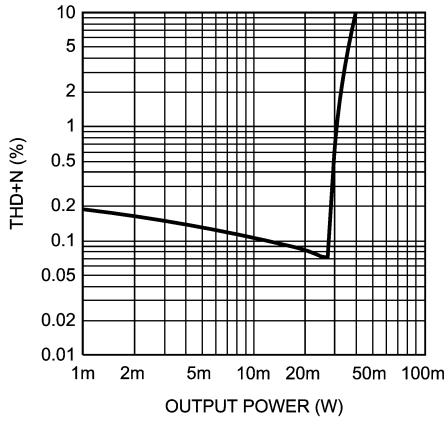


201697H4

Typical Performance Characteristics (Continued)

THD+N vs Output Power

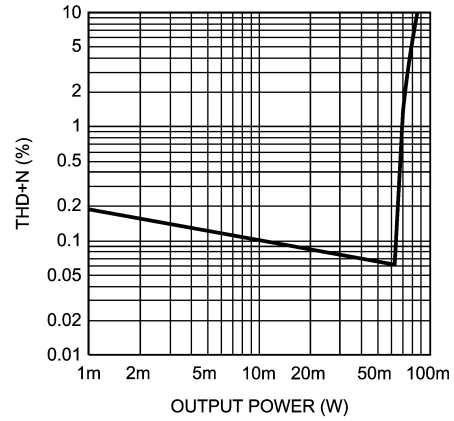
$V_{DD} = 2.5V, R_L = 16\Omega$
OCL



20169758

THD+N vs Output Power

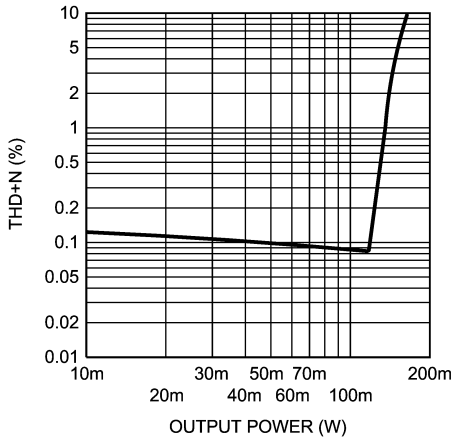
$V_{DD} = 3.6V, R_L = 16\Omega$
OCL



20169759

THD+N vs Output Power

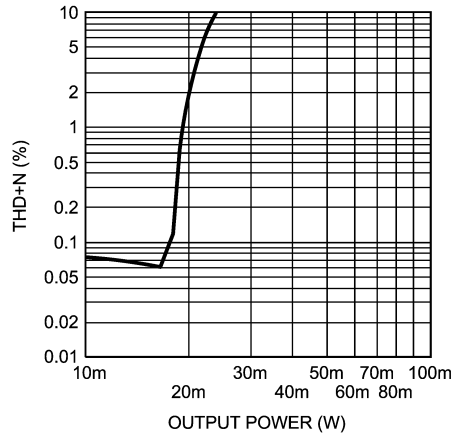
$V_{DD} = 5.0V, R_L = 16\Omega$
OCL



20169760

THD+N vs Output Power

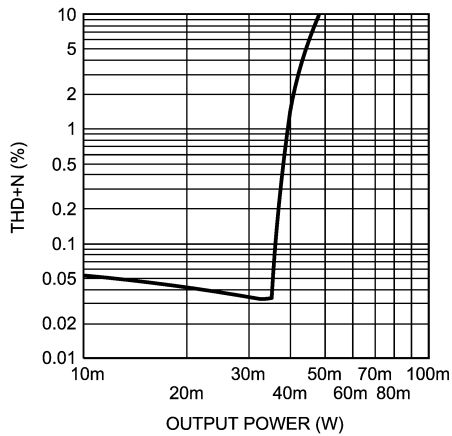
$V_{DD} = 2.5V, R_L = 32\Omega$
OCL



20169761

THD+N vs Output Power

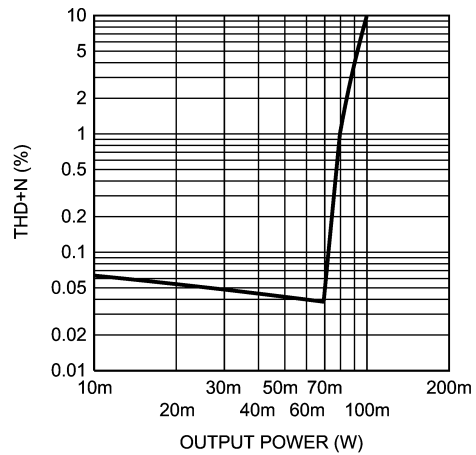
$V_{DD} = 3.6V, R_L = 32\Omega$
OCL



20169762

THD+N vs Output Power

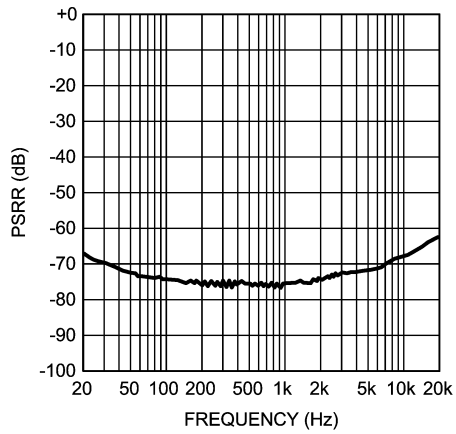
$V_{DD} = 5.0V, R_L = 32\Omega$
OCL



20169763

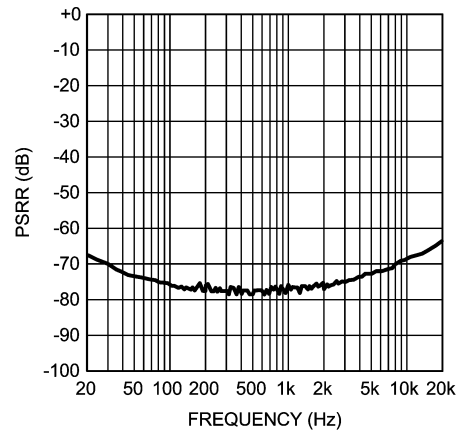
Typical Performance Characteristics (Continued)

PSRR vs Frequency
 $V_{DD} = 2.5V$, $R_L = 16\Omega$
 $V_{RIPPLE} = 200mV_{pp}$, OCL



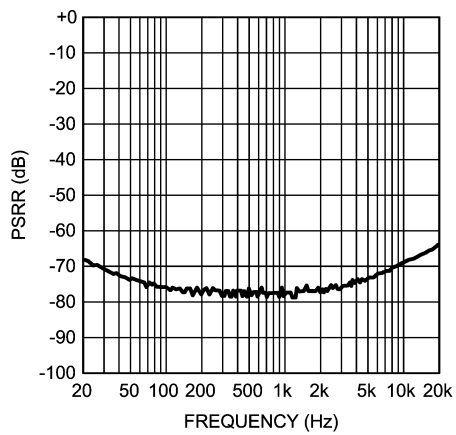
20169776

PSRR vs Frequency
 $V_{DD} = 3.6V$, $R_L = 16\Omega$
 $V_{RIPPLE} = 200mV_{pp}$, OCL



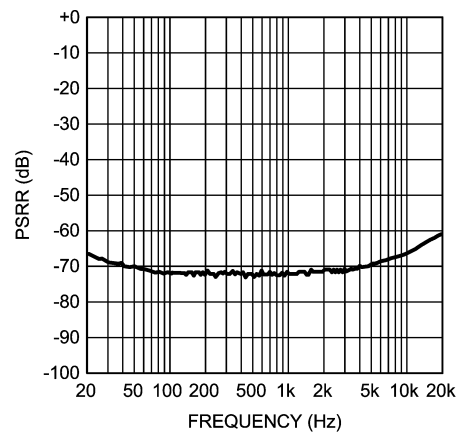
201697H5

PSRR vs Frequency
 $V_{DD} = 5.0V$, $R_L = 16\Omega$
 $V_{RIPPLE} = 200mV_{pp}$, OCL



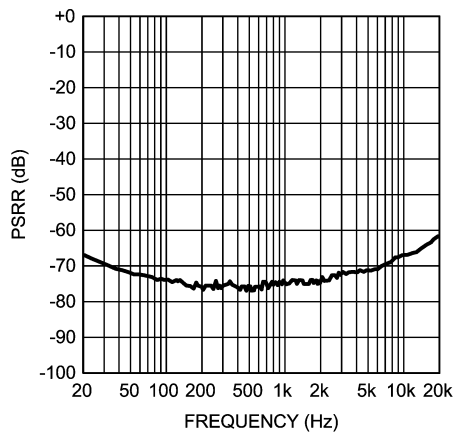
201697H6

PSRR vs Frequency
 $V_{DD} = 2.5V$, $R_L = 32\Omega$
 $V_{RIPPLE} = 200mV_{pp}$, OCL



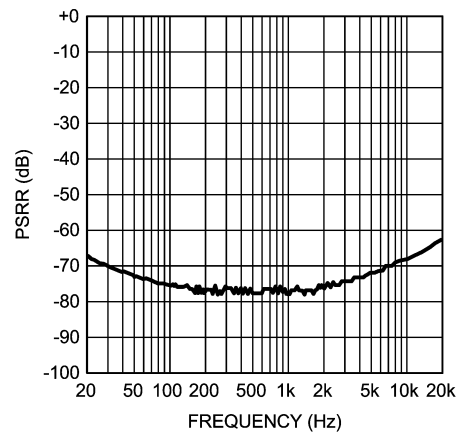
201697H7

PSRR vs Frequency
 $V_{DD} = 3.6V$, $R_L = 32\Omega$
 $V_{RIPPLE} = 200mV_{pp}$, OCL



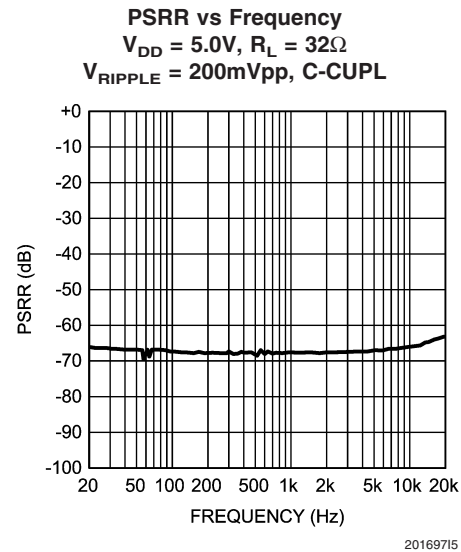
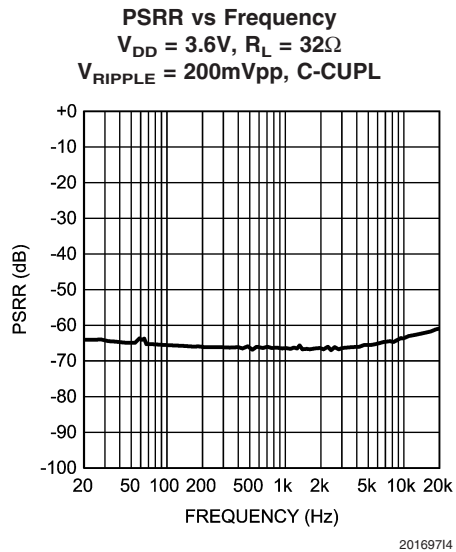
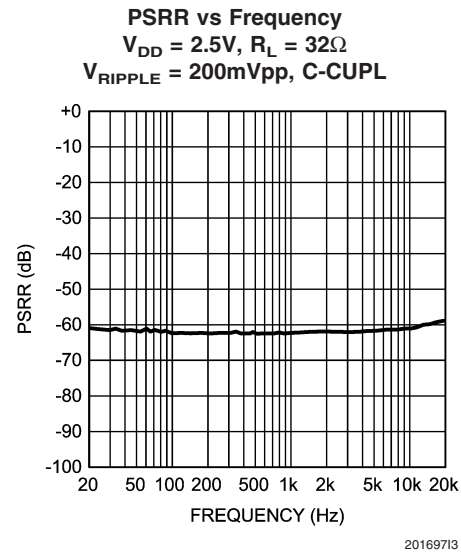
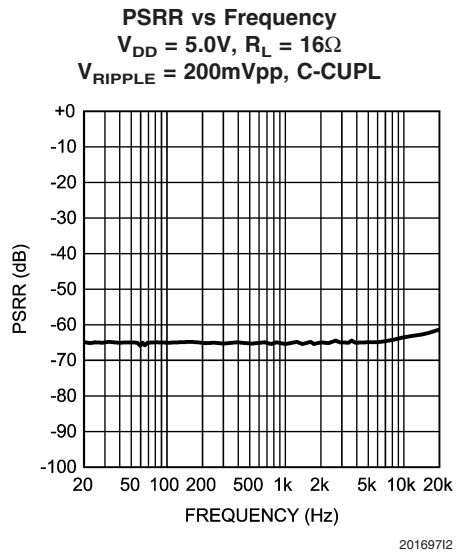
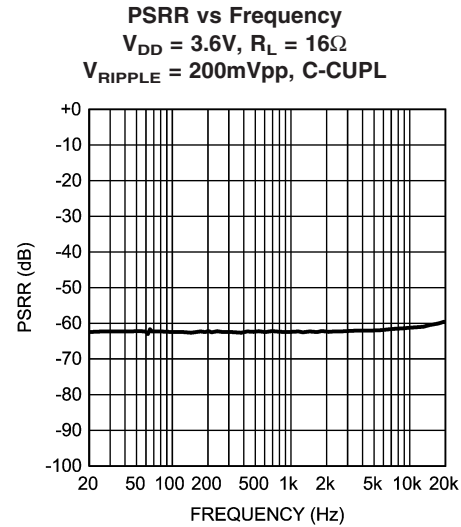
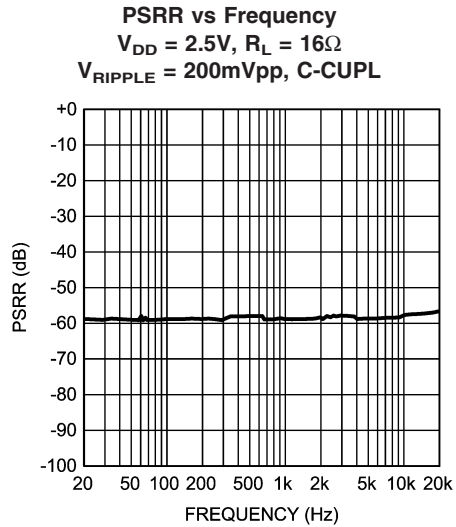
201697H8

PSRR vs Frequency
 $V_{DD} = 5.0V$, $R_L = 32\Omega$
 $V_{RIPPLE} = 200mV_{pp}$, OCL



201697H9

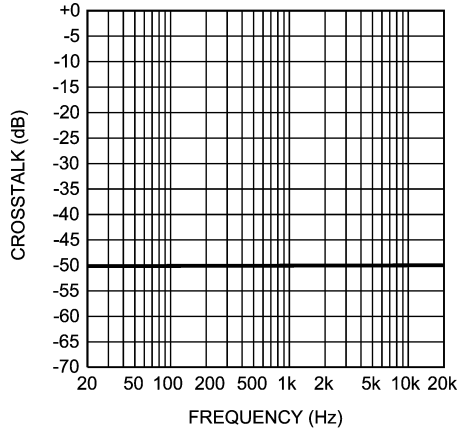
Typical Performance Characteristics (Continued)



Typical Performance Characteristics (Continued)

Crosstalk vs Frequency

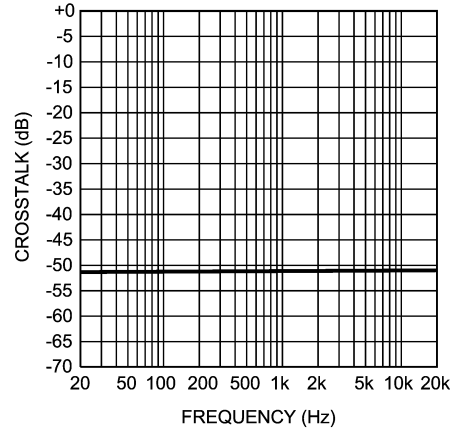
$V_{DD} = 2.5V$, $R_L = 16\Omega$
 $P_{OUT} = 20mW$, OCL



201697I6

Crosstalk vs Frequency

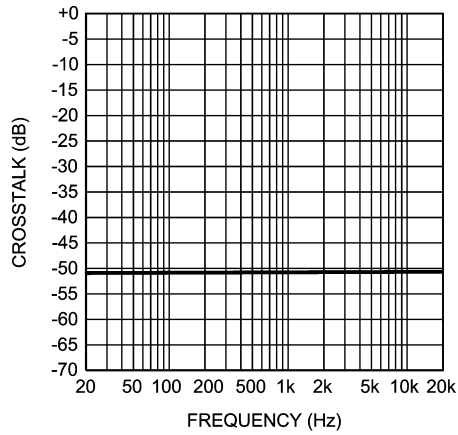
$V_{DD} = 3.6V$, $R_L = 16\Omega$
 $P_{OUT} = 40mW$, OCL



201697G8

Crosstalk vs Frequency

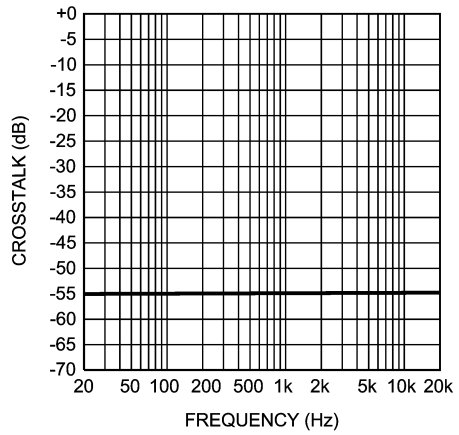
$V_{DD} = 5.0V$, $R_L = 16\Omega$
 $P_{OUT} = 40mW$, OCL



201697G9

Crosstalk vs Frequency

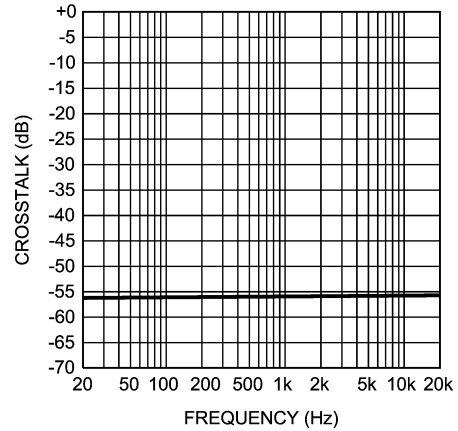
$V_{DD} = 2.5V$, $R_L = 32\Omega$
 $P_{OUT} = 20mW$, OCL



201697H0

Crosstalk vs Frequency

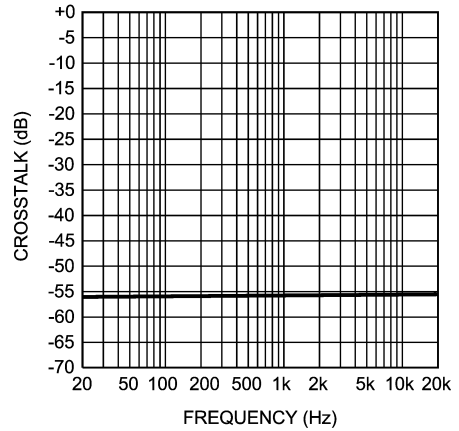
$V_{DD} = 3.6V$, $R_L = 32\Omega$
 $P_{OUT} = 40mW$, OCL



201697H1

Crosstalk vs Frequency

$V_{DD} = 5.0V$, $R_L = 32\Omega$
 $P_{OUT} = 50mW$, OCL

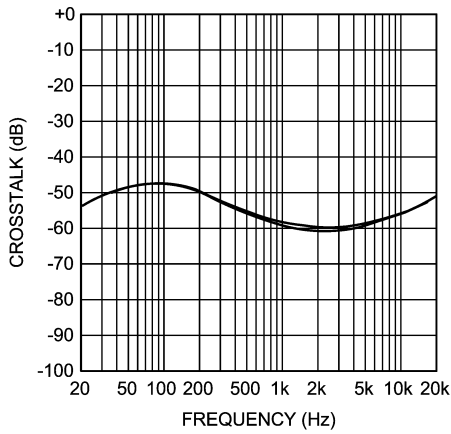


201697H2

Typical Performance Characteristics (Continued)

Crosstalk vs Frequency

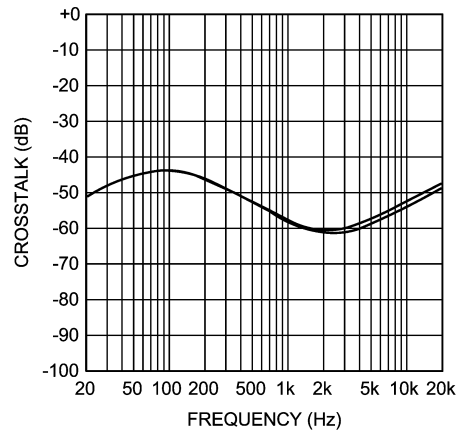
$V_{DD} = 2.5V, R_L = 16\Omega$
 $P_{OUT} = 20mW, C-CUPL$



201697D7

Crosstalk vs Frequency

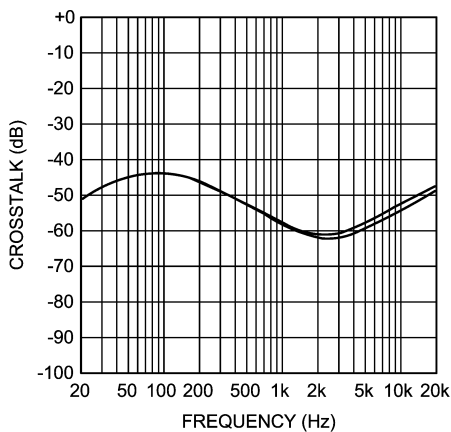
$V_{DD} = 3.6V, R_L = 16\Omega$
 $P_{OUT} = 50mW, C-CUPL$



201697D8

Crosstalk vs Frequency

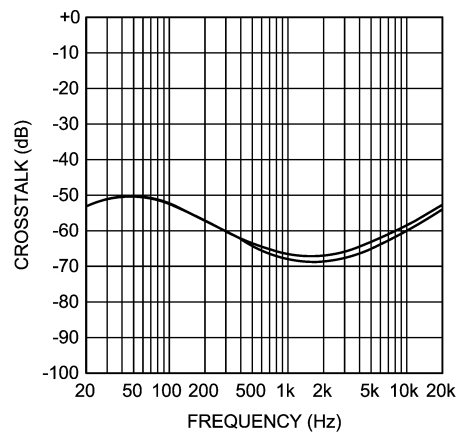
$V_{DD} = 5.0V, R_L = 16\Omega$
 $P_{OUT} = 50mW, C-CUPL$



201697D9

Crosstalk vs Frequency

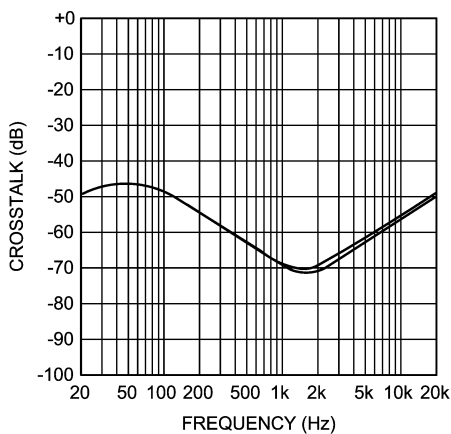
$V_{DD} = 2.5V, R_L = 32\Omega$
 $P_{OUT} = 20mW, C-CUPL$



201697E0

Crosstalk vs Frequency

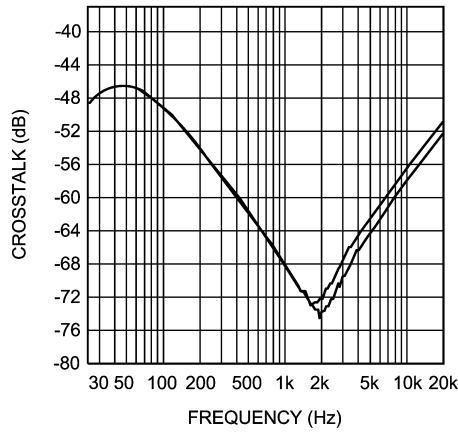
$V_{DD} = 3.6V, R_L = 32\Omega$
 $P_{OUT} = 50mW, C-CUPL$



201697E1

Crosstalk vs Frequency

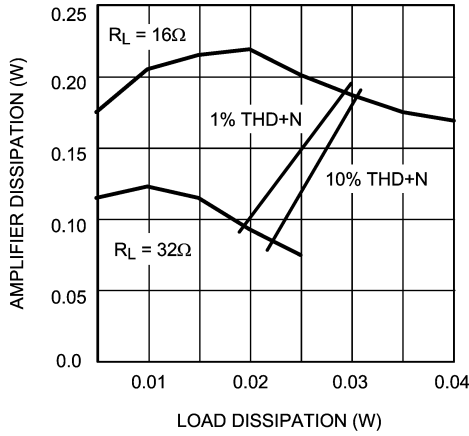
$V_{DD} = 5.0V, R_L = 32\Omega$
 $P_{OUT} = 50mW, C-CUPL$



201697E2

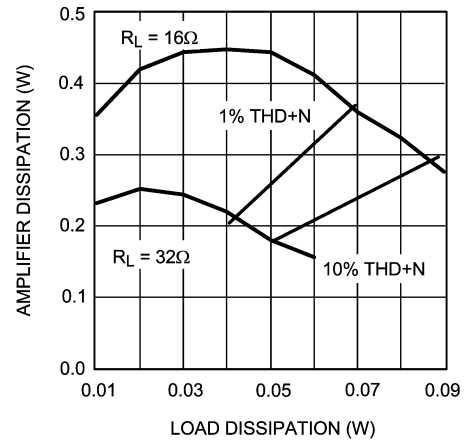
Typical Performance Characteristics (Continued)

Load Dissipation vs Amplifier Dissipation
 $V_{DD} = 2.5V$, C-CUPL



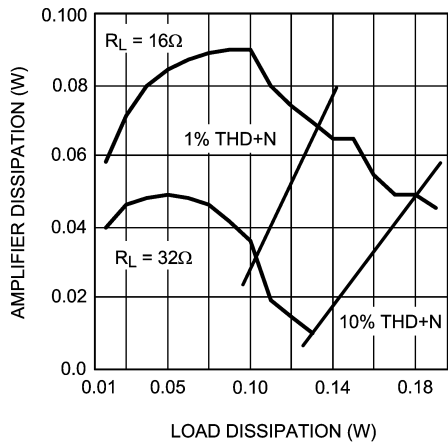
20169755

Load Dissipation vs Amplifier Dissipation
 $V_{DD} = 3.6V$, C-CUPL



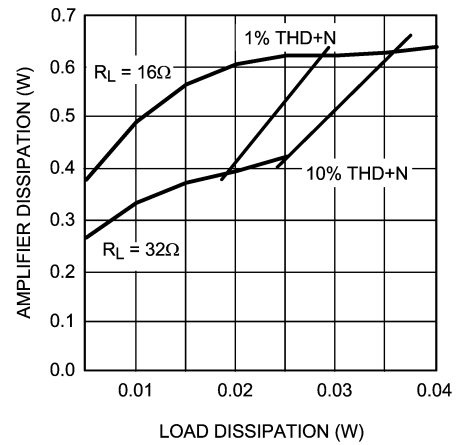
20169756

Load Dissipation vs Amplifier Dissipation
 $V_{DD} = 5.0V$, C-CUPL



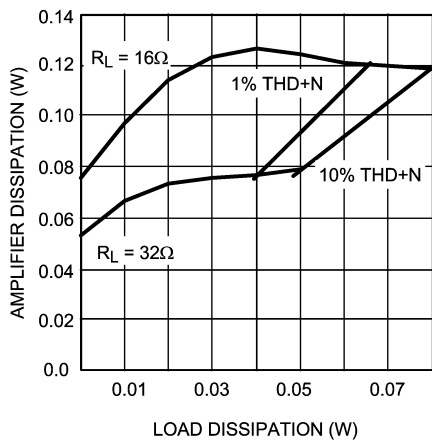
20169757

Load Dissipation vs Amplifier Dissipation
 $V_{DD} = 2.5V$, OCL



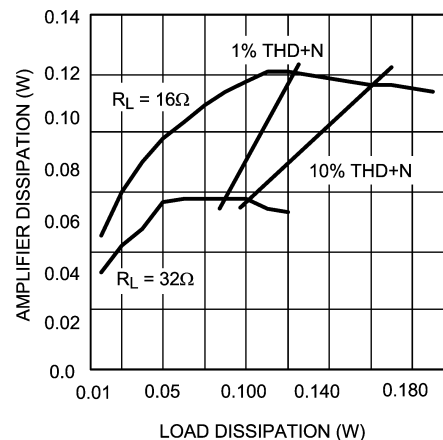
20169738

Load Dissipation vs Amplifier Dissipation
 $V_{DD} = 3.6V$, OCL



20169739

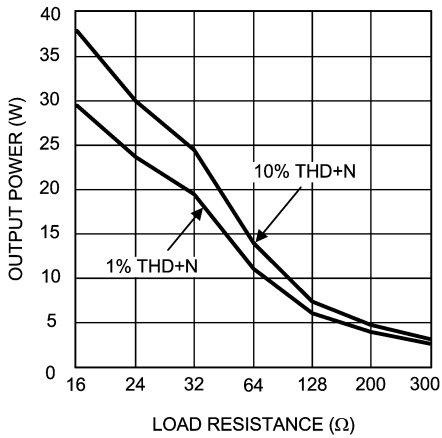
Load Dissipation vs Amplifier Dissipation
 $V_{DD} = 5.0V$, OCL



20169740

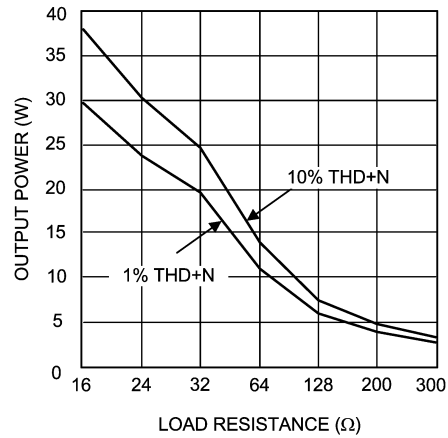
Typical Performance Characteristics (Continued)

Output Power vs Load Resistance
 $V_{DD} = 2.5V, C-CUPL$



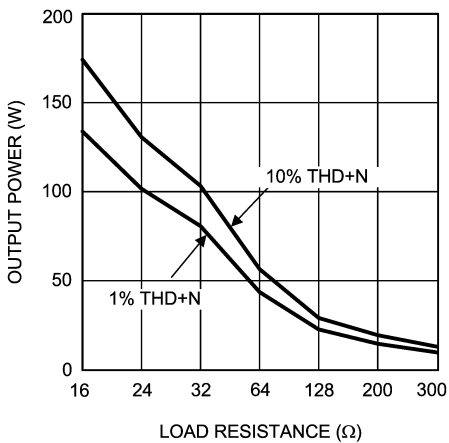
20169741

Output Power vs Load Resistance
 $V_{DD} = 3.6V, C-CUPL$



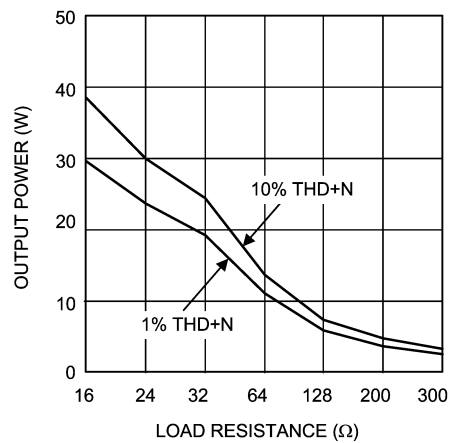
20169742

Output Power vs Load Resistance
 $V_{DD} = 5.0V, C-CUPL$



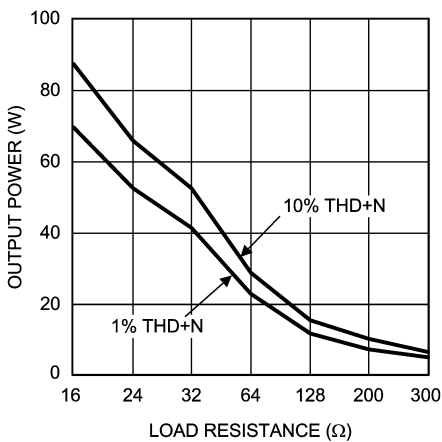
20169743

Output Power vs Load Resistance
 $V_{DD} = 2.5V, OCL$



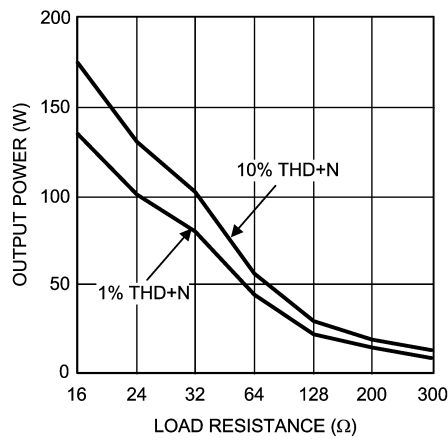
20169744

Output Power vs Load Resistance
 $V_{DD} = 3.6V, OCL$



20169745

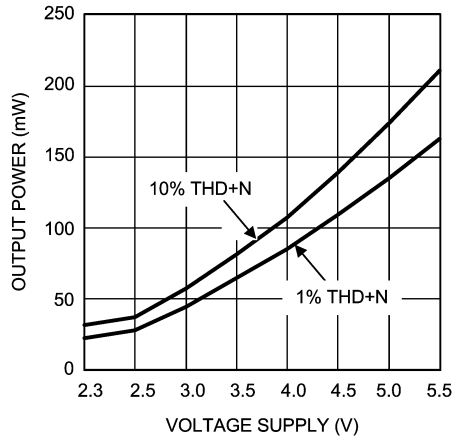
Output Power vs Load Resistance
 $V_{DD} = 5.0V, OCL$



20169746

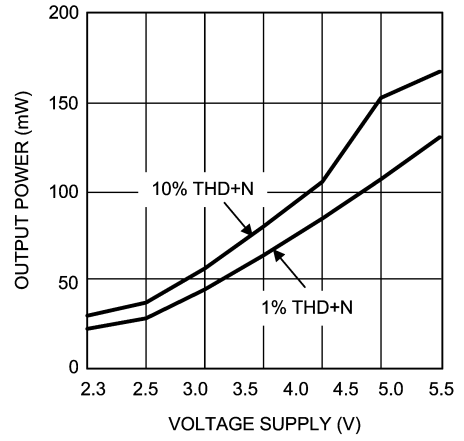
Typical Performance Characteristics (Continued)

Output Power vs Supply Voltage
 $R_L = 16\Omega$, C-CUPL



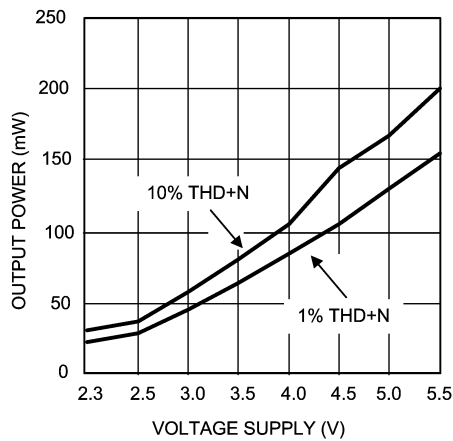
20169747

Output Power vs Supply Voltage
 $R_L = 32\Omega$, C-CUPL



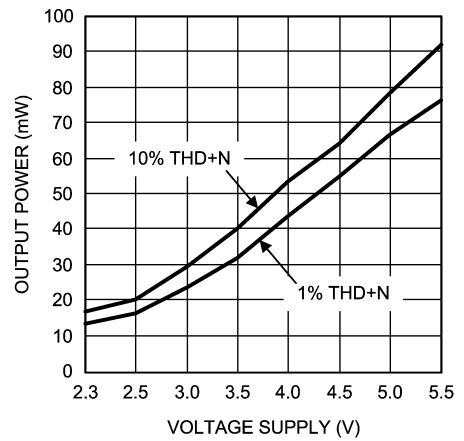
20169748

Output Power vs Supply Voltage
 $R_L = 16\Omega$, OCL



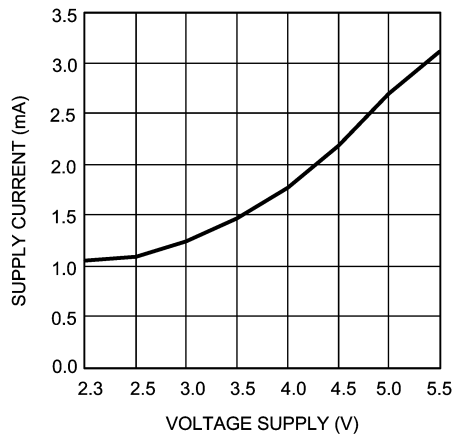
20169749

Output Power vs Supply Voltage
 $R_L = 32\Omega$, OCL



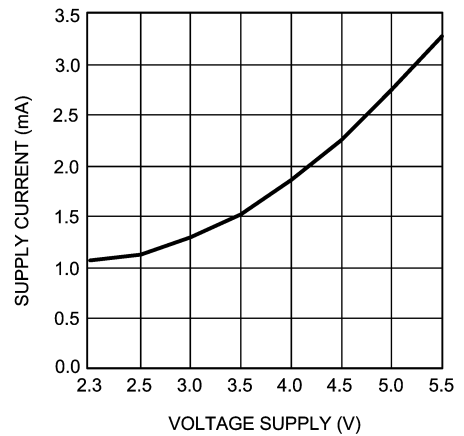
20169750

Supply Current vs Supply Voltage
 $R_L = 16\Omega$, C-CUPL



20169751

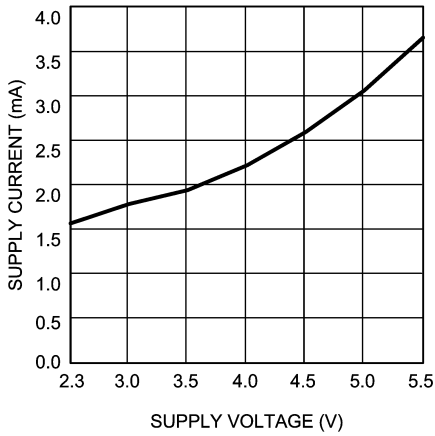
Supply Current vs Supply Voltage
 $R_L = 32\Omega$, C-CUPL



20169752

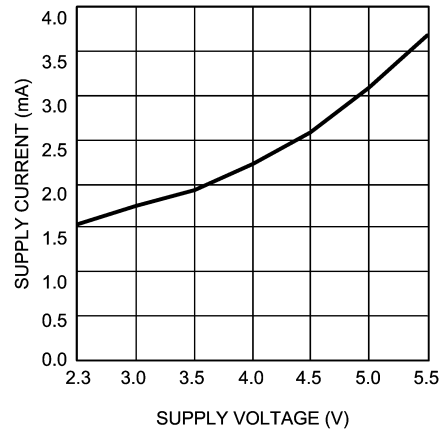
Typical Performance Characteristics (Continued)

Supply Current vs Supply Voltage
 $R_L = 16\Omega$, OCL



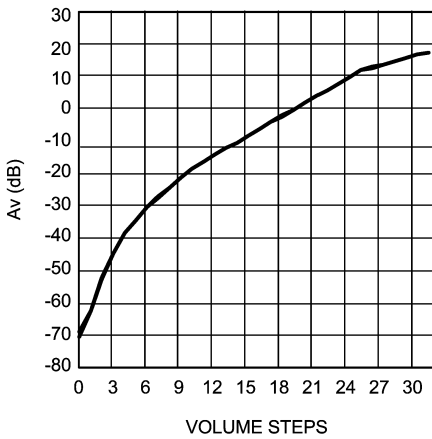
20169753

Supply Current vs Supply Voltage
 $R_L = 32\Omega$, OCL



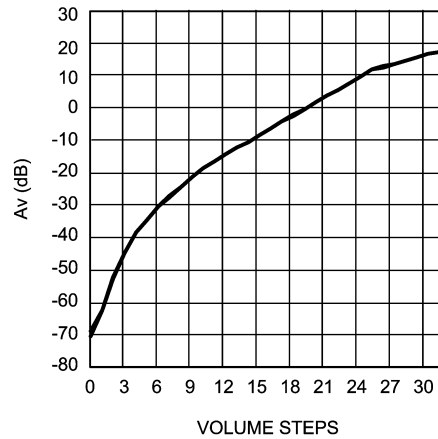
20169754

Gain vs Volume Steps
 $V_{CC} = 2.5V$, $R_L = 16\Omega$, OCL



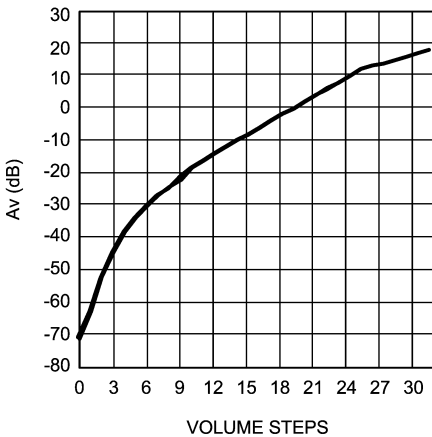
201697F7

Gain vs Volume Steps
 $V_{CC} = 3.6V$, $R_L = 16\Omega$, OCL



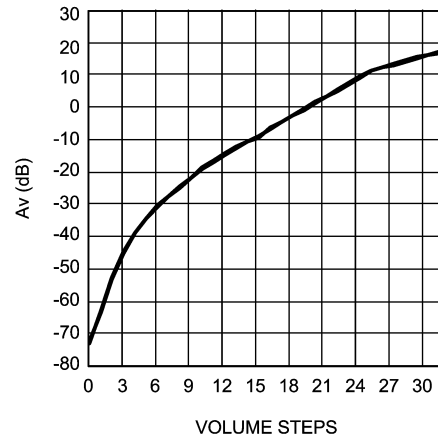
201697F5

Gain vs Volume Steps
 $V_{CC} = 5V$, $R_L = 16\Omega$, OCL



201697G4

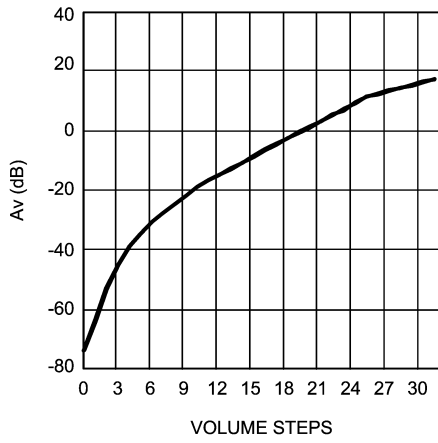
Gain vs Volume Steps
 $V_{CC} = 2.5V$, $R_L = 16\Omega$, C-CUPL



201697F6

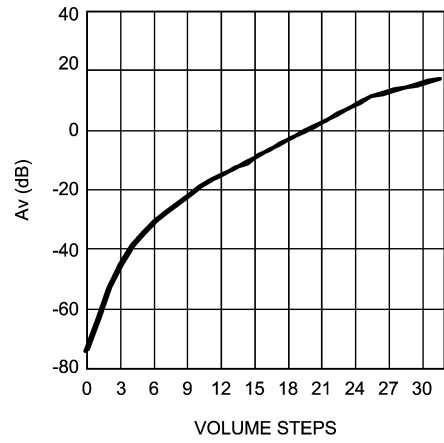
Typical Performance Characteristics (Continued)

Gain vs Volume Steps
 $V_{CC} = 3.6V, R_L = 16\Omega, C-CUPL$



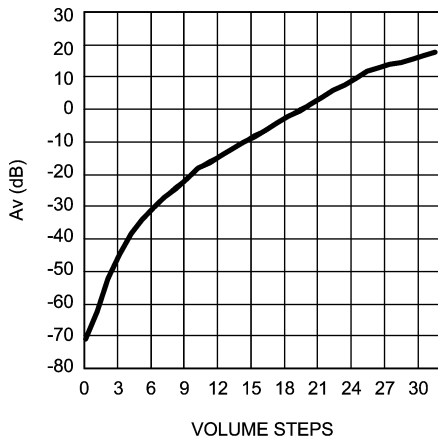
201697G0

Gain vs Volume Steps
 $V_{CC} = 5V, R_L = 16\Omega, C-CUPL$



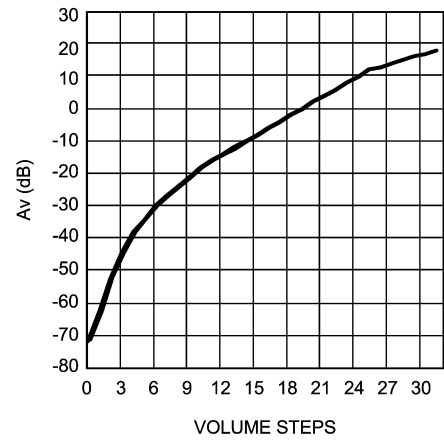
201697G3

Gain vs Volume Steps
 $V_{CC} = 2.5V, R_L = 32\Omega, OCL$



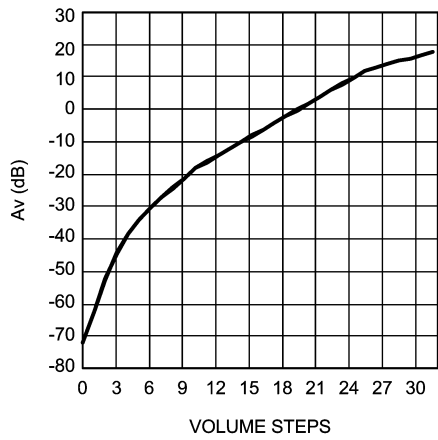
201697F9

Gain vs Volume Steps
 $V_{CC} = 3.6V, R_L = 32\Omega, OCL$



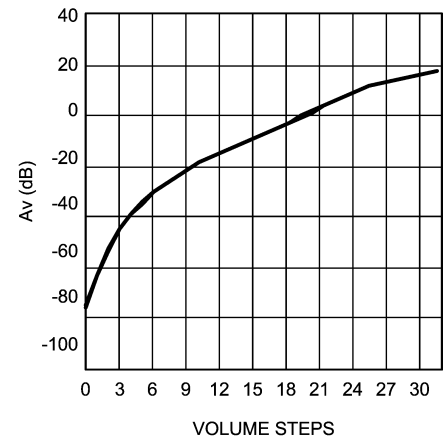
201697G2

Gain vs Volume Steps
 $V_{CC} = 5V, R_L = 32\Omega, OCL$



201697G6

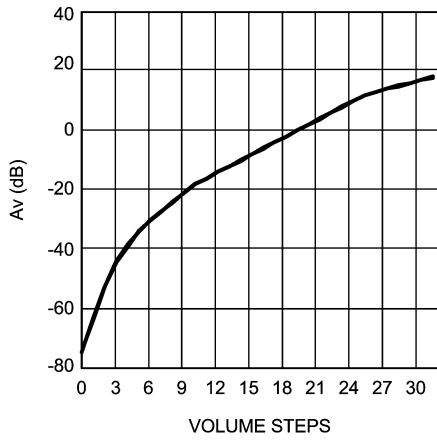
Gain vs Volume Steps
 $V_{CC} = 2.5V, R_L = 32\Omega, C-CUPL$



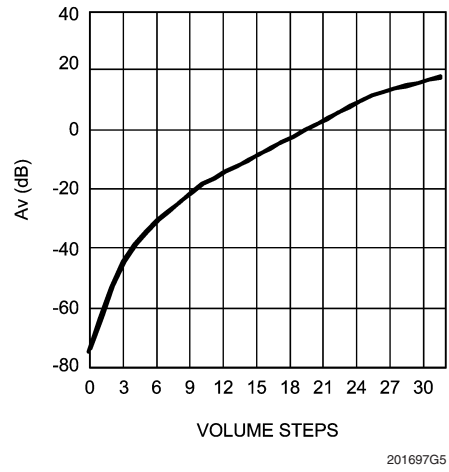
201697F8

Typical Performance Characteristics (Continued)

Gain vs Volume Steps
 $V_{CC} = 3.6V, R_L = 32\Omega, C-CUPL$



Gain vs Volume Steps
 $V_{CC} = 5V, R_L = 32\Omega, C-CUPL$



Application Information

AMPLIFIER CONFIGURATION EXPLANATION

As shown in Figure 1, the LM4985 has three internal power amplifiers. Two of the amplifiers which amplify signals applied to their inputs, have internally configurable gain. The remaining third amplifier provides both half-supply output bias and AC ground return.

Loads, such as a headphone speaker, are connected between OUT1 and CNTGND or OUT2 and CNTGND. This configuration does not require an output coupling capacitor. The classical single-ended amplifier configuration, where one side of the load is connected to ground, requires large, expensive output coupling capacitors.

A configuration such as the one used in the LM4985 has a major advantage over single supply, single-ended amplifiers. Since the outputs OUT1, OUT2, and CNTGND are all biased at $1/2 V_{DD}$, no net DC voltage exists across each load. This eliminates the need for output coupling capacitors which are required in a single-supply, single-ended amplifier configuration. Without output coupling capacitors in a typical single-supply, single-ended amplifier, the bias voltage is placed across the load resulting in both increased internal IC power dissipation and possible loudspeaker damage.

The LM4985 eliminates these output coupling capacitors when operating in Output Capacitor-less (OCL) mode. Unless shorted to ground, VoC is internally configured to apply a $1/2 V_{DD}$ bias voltage to a stereo headphone jack's sleeve. This voltage matches the bias voltage present on VoA and VoB outputs that drive the headphones. The headphones operate in a manner similar to a bridge-tied load (BTL). Because the same DC voltage is applied to both headphone speaker terminals this results in no net DC current flow through the speaker. AC current flows through a headphone speaker as an audio signal's output amplitude increases on the speaker's terminal.

The headphone jack's sleeve is not connected to circuit ground when used in OCL mode. Using the headphone output jack as a line-level output will place the LM4985's $1/2 V_{DD}$ bias voltage on a plug's sleeve connection. This presents no difficulty when the external equipment uses capacitively coupled inputs. For the very small minority of equipment that is DC coupled, the LM4985 monitors the current supplied by the amplifier that drives the headphone jack's sleeve. If this current exceeds 500mA_{PEAK} , the amplifier is shutdown, protecting the LM4985 and the external equipment.

POWER DISSIPATION

Power dissipation is a major concern when using any power amplifier. When operating in capacitor-coupled mode (C-CUPL), Equation 1 states the maximum power dissipation point for a single-ended amplifier operating at a given supply voltage and driving a specified output load.

$$P_{\text{DMAX}} = 2(V_{\text{DD}})^2 / (2\pi^2 R_L) \quad (1)$$

When operating in the OCL mode, the LM4985's three operational amplifiers produce a maximum power dissipation given in Equation 2:

$$P_{\text{DMAX}} = [2(V_{\text{DD}})^2 / (2\pi^2 R_L)] + [V_{\text{DD}}^2 / (4\pi R_L)] \quad (2)$$

The maximum power dissipation point obtained from Equation 1 or Equation 2 must not be greater than the power dissipation that results from Equation 3:

$$P_{\text{DMAX}} = (T_{\text{JMAX}} - T_A) / \theta_{\text{JA}} \quad (3)$$

For package TMD12AAA, $\theta_{\text{JA}} = 190^\circ\text{C/W}$. $T_{\text{JMAX}} = 150^\circ\text{C}$ for the LM4985. Depending on the ambient temperature, T_A , of the system surroundings, Equation 3 can be used to find the maximum internal power dissipation supported by the IC packaging. If the result of Equation 2 is greater than that of Equation 3, then either the supply voltage must be decreased, the load impedance increased or T_A reduced.

For a typical application using a 3.6V power supply, with a 32Ω load, the maximum ambient temperature possible without violating the maximum junction temperature is approximately 144°C provided that device operation is around the maximum power dissipation point. Thus, for typical applications, power dissipation is not an issue. Power dissipation is a function of output power and thus, if typical operation is not around the maximum power dissipation point, the ambient temperature may be increased accordingly. Refer to the Typical Performance Characteristics curves for power dissipation information for lower output powers.

POWER SUPPLY BYPASSING

As with any amplifier, proper supply bypassing is important for low noise performance and high power supply rejection. The capacitor location on the power supply pins should be as close to the device as possible.

Typical applications employ a regulator with $10\mu\text{F}$ tantalum or electrolytic capacitor and a ceramic bypass capacitor which aid in supply stability. This does not eliminate the need for bypassing the supply nodes of the LM4985. A bypass capacitor value in the range of $0.1\mu\text{F}$ to $1\mu\text{F}$ is recommended for C_S .

MICRO POWER SHUTDOWN

The LM4985's micropower shutdown is activated or deactivated through its I²C digital interface. Please refer to Table 1 for the I²C Address, Register Select, and Mode Control registers. Each amplifier within the LM4985 can be shutdown individually.

Please observe the following protocol when placing an individual amplifier channel in shutdown while the other channel remains active. The protocol requires activating both channels' shutdown simultaneously, then deactivating the shutdown of the channel whose output is desired (or leaving the desired channel in shutdown mode). Also, when operating in the C-CUPL mode, a short delay time is required between activating one channel after placing both channels in shutdown. If the user finds that both channels activate when only one was chosen, increase the delay.

SELECTION OF INPUT CAPACITOR SIZE

Amplifying the lowest audio frequencies requires a high value input coupling capacitor, C_i . A high value capacitor can be expensive and may compromise space efficiency in portable designs. In many cases, however, the headphones used in portable systems have little ability to reproduce signals below 60Hz. Applications using headphones with this limited frequency response reap little improvement by using a high value input capacitor.

In addition to system cost and size, turn on time is affected by the size of the input coupling capacitor C_i . A larger input

Application Information (Continued)

coupling capacitor requires more charge to reach its quiescent DC voltage. This charge comes from the output via the feedback. Thus, by minimizing the capacitor size based on necessary low frequency response, turn-on time can be minimized. A small value of C_i (in the range of $0.22\mu\text{F}$ to $0.68\mu\text{F}$), is recommended.

MAXIMIZING OCL MODE CHANNEL-to-CHANNEL SEPARATION

The OCL mode AC ground return (CNT_GND pin) is shared by both amplifiers. As such, any resistance between the CNT_GND pin and the load will create a voltage divider with respect to the load resistance. In a typical circuit, the amount of CNT_GND resistance can be very small, but still significant. It is significant because of the relatively low load impedances for which the LM4985 was designed to drive: 16Ω

to 32Ω . The ratio of this voltage divider will determine the magnitude of any residual signal present at the CNT_GND pin. It is this residual signal that leads to channel-to-channel separation (crosstalk) degradation.

For example, for a 60dB channel-to-channel separation while driving a 16Ω load, the resistance between the LM4985's CNT_GND pin and the load must be less than $16\text{m}\Omega$. This is achieved by ensuring that the trace that connects the CNT_GND pin to the headphone jack sleeve should be as short and massive as possible, given the physical constraints of any specific printed circuit board layout and design.

DEMONSTRATION BOARD AND PCB LAYOUT

Information concerning PCB layout considerations and demonstration board use and performance is found in Application Note AN-1452.

I²C Control Register

Table 1 shows the actions that are implemented by manipulating the bits within the two internal I²C control registers.

Table 1. LM4985 I²C Control Register Addressing and Data Format Chart

LM4985 I ² C Control Register Addressing and Data Chart									
I ² C Address	A6	A5	A4	A3	A2	A1	A0	Function	
	1	1	0	0	1	1	A0		
Register Select	D7	D6	D5	D4	D3	D2	RS1	RS0	
	0	0	0	0	0	0	0	0	Read and write the mode control register
	0	0	0	0	0	0	0	1	Read and write the volume control register
Mode Control Register	D7	D6	D5	D4	D3	D2	D1	D0	
		WT1	WT0	PHG	SDCH1	SDCH2	CHSEL1	CHSEL2	
	0	X	X	X	X	X	X	X	D7 must always be set to 0
	–	0	0	X	X	X	X	X	Wake-up time: 80ms (OCL), 250ms (C-CUPL)
	–	0	1	X	X	X	X	X	Wake-up time: 110ms (OCL), 450ms (C-CUPL)
	–	1	0	X	X	X	X	X	Wake-up time: 170ms (OCL), 850ms (C-CUPL)
	–	1	1	X	X	X	X	X	Wake-up time: 290ms (OCL), 1650ms (C-CUPL)
	–	X	X	1	X	X	X	X	Output capacitor-less mode active
	–	X	X	0	X	X	X	X	Output capacitor-less mode inactive
	–	X	X	X	0	0	X	X	Amplifier's SHUTDOWN mode active
	–	X	X	X	0	1	X	X	Illegal mode
	–	X	X	X	1	0	X	X	Illegal mode
	–	X	X	X	1	1	X	X	Amplifier's SHUTDOWN mode inactive
	–	X	X	X	X	X	0	02	Amplifier's Chan. 1 is Input 1 , Chan 2. is Input 2
	–	X	X	X	X	X	0	1	Amplifier's Chan. 1 is Input 1 , Chan 2. is Input 1
	–	X	X	X	X	X	1	0	Amplifier's Chan. 1 is Input 2 , Chan 2. is Input 2
–	X	X	X	X	X	1	1	Amplifier's Chan. 1 is Input 2 , Chan 2. is Input 1	

Volume Control Settings Binary Values

The minimum volume setting is set to -76dB when 00000 is loaded into the volume control register. Incrementing the volume control register in binary fashion increases the volume control setting, reaching full scale at 11111. Table C1 shows the value of the gain for each of the 32 binary volume control settings.

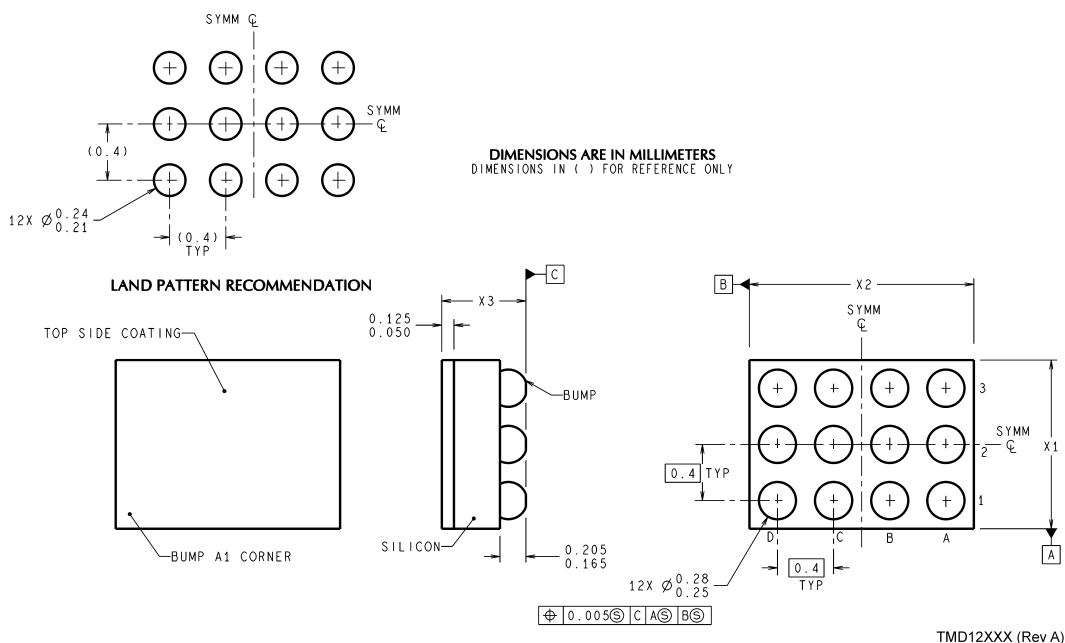
Table C1. Binary Values for the Different Volume Control Gain Settings

Gain	B4	B3	B2	B1	B0
18	1	1	1	1	1
17	1	1	1	1	0
16	1	1	1	0	1
15	1	1	1	0	0
14	1	1	0	1	1
13	1	1	0	1	0
12	1	1	0	0	1
10	1	1	0	0	0
8	1	0	1	1	1
6	1	0	1	1	0
4	1	0	1	0	1
2	1	0	1	0	0
0	1	0	0	1	1
-2	1	0	0	1	0
-4	1	0	0	0	1
-6	1	0	0	0	0
-8	0	1	1	1	1
-10	0	1	1	1	0
-12	0	1	1	0	1
-14	0	1	1	0	0
-16	0	1	0	1	1
-18	0	1	0	1	0
-21	0	1	0	0	1
-24	0	1	0	0	0
-27	0	0	1	1	1
-30	0	0	1	1	0
-34	0	0	1	0	1
-38	0	0	1	0	0
-44	0	0	0	1	1
-52	0	0	0	1	0
-62	0	0	0	0	1
-76	0	0	0	0	0

Revision History

Rev	Date	Description
1.0	05/17/06	Initial WEB release.

Physical Dimensions inches (millimeters) unless otherwise noted



micro SMD
Order Number LM4985TM
NS Package Number TMD12AAA
X₁ = 1.215mm ± 0.03mm X₂ = 1.615mm ± 0.03mm X₃ = 0.600mm ± 0.075mm

National does not assume any responsibility for use of any circuitry described, no circuit patent licenses are implied and National reserves the right at any time without notice to change said circuitry and specifications.

For the most current product information visit us at www.national.com.

LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT AND GENERAL COUNSEL OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

BANNED SUBSTANCE COMPLIANCE

National Semiconductor follows the provisions of the Product Stewardship Guide for Customers (CSP-9-111C2) and Banned Substances and Materials of Interest Specification (CSP-9-111S2) for regulatory environmental compliance. Details may be found at: www.national.com/quality/green.

Lead free products are RoHS compliant.



National Semiconductor
Americas Customer
Support Center
 Email: new.feedback@nsc.com
 Tel: 1-800-272-9959

National Semiconductor
Europe Customer Support Center
 Fax: +49 (0) 180-530 85 86
 Email: europe.support@nsc.com
 Deutsch Tel: +49 (0) 69 9508 6208
 English Tel: +44 (0) 870 24 0 2171
 Français Tel: +33 (0) 1 41 91 8790

National Semiconductor
Asia Pacific Customer
Support Center
 Email: ap.support@nsc.com

National Semiconductor
Japan Customer Support Center
 Fax: 81-3-5639-7507
 Email: jpn.feedback@nsc.com
 Tel: 81-3-5639-7560

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All products are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

TI assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using TI components. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any TI patent right, copyright, mask work right, or other TI intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information published by TI regarding third-party products or services does not constitute a license from TI to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. Reproduction of this information with alteration is an unfair and deceptive business practice. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

TI products are not authorized for use in safety-critical applications (such as life support) where a failure of the TI product would reasonably be expected to cause severe personal injury or death, unless officers of the parties have executed an agreement specifically governing such use. Buyers represent that they have all necessary expertise in the safety and regulatory ramifications of their applications, and acknowledge and agree that they are solely responsible for all legal, regulatory and safety-related requirements concerning their products and any use of TI products in such safety-critical applications, notwithstanding any applications-related information or support that may be provided by TI. Further, Buyers must fully indemnify TI and its representatives against any damages arising out of the use of TI products in such safety-critical applications.

TI products are neither designed nor intended for use in military/aerospace applications or environments unless the TI products are specifically designated by TI as military-grade or "enhanced plastic." Only products designated by TI as military-grade meet military specifications. Buyers acknowledge and agree that any such use of TI products which TI has not designated as military-grade is solely at the Buyer's risk, and that they are solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI products are neither designed nor intended for use in automotive applications or environments unless the specific TI products are designated by TI as compliant with ISO/TS 16949 requirements. Buyers acknowledge and agree that, if they use any non-designated products in automotive applications, TI will not be responsible for any failure to meet such requirements.

Following are URLs where you can obtain information on other Texas Instruments products and application solutions:

Products

Audio	www.ti.com/audio
Amplifiers	amplifier.ti.com
Data Converters	dataconverter.ti.com
DLP® Products	www.dlp.com
DSP	dsp.ti.com
Clocks and Timers	www.ti.com/clocks
Interface	interface.ti.com
Logic	logic.ti.com
Power Mgmt	power.ti.com
Microcontrollers	microcontroller.ti.com
RFID	www.ti-rfid.com
OMAP Mobile Processors	www.ti.com/omap
Wireless Connectivity	www.ti.com/wirelessconnectivity

Applications

Communications and Telecom	www.ti.com/communications
Computers and Peripherals	www.ti.com/computers
Consumer Electronics	www.ti.com/consumer-apps
Energy and Lighting	www.ti.com/energy
Industrial	www.ti.com/industrial
Medical	www.ti.com/medical
Security	www.ti.com/security
Space, Avionics and Defense	www.ti.com/space-avionics-defense
Transportation and Automotive	www.ti.com/automotive
Video and Imaging	www.ti.com/video

TI E2E Community Home Page

e2e.ti.com

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265
Copyright © 2011, Texas Instruments Incorporated